# Memristor for bio-inspired computing: <br> nonlinear circuits and mathematical models 

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## Outline

- Motivation
- Memristor modeling and nonlinear dynamics
- Model comparison
- Influence on initial conditions
- the SIMPLEST (passive) memristor circuit
- Conclusions

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## Neuromorphic circuits

- The Human Brain Project in EU plans to use a supercomputer to recreate everything known about the human brain - a hugely ambitious goal!
- Leading neuroscientists in the US are now focussed on understanding how the brain works through the Brain Activity Map (BAM) project, but it's difficult to peer deeply enough into a brain to map the activity of every neuron. Because zebrafish embryos are transparent, the task is easier.
- Understand how neurons that make up the brain carry out their functions.


## Neuromorphic circuits Technology Review

So the race is on to develop a different kind of chip that more accurately mimics the way the brain works. So-called neuromorphic chips must be built from devices that behave like neurons - in other words they transmit and respond to information sent in spikes rather than in a continously varying voltage.

One reason the brain is so power efficient is that neural spikes charge only a small fraction of a neuron as they travel. By contrast, conventional chips keep each and every transmission line at a certain voltage all the time.

Clearly, recent advancements in memristor technology and spintronics are making possible entirely new ways to design chips. However, there is a long way to go before synthetic systems can begin to match the capability of natural ones.

# Beyond Moore's Law 

## Science and Engineering

 Beyond Moore's Law| Device | Entity | Properties |  |  |
| :--- | :--- | :--- | :---: | :---: |
|  |  | Control Variable | State Variable | Output Variable |
| FET - Novel Materials (III-V, <br> Ge, carbon-based, etc.) | Electron | Charge | Charge | Charge |
| SpinFET | Electron |  | Spin | Charge |
| Spin-Torque | Electron | Spin | Spin | Charge |
| Spin-Wave | Electron | Spin Waves | Spin | Charge <br> Photon |
| Tunneling Transistor | Electron | Charge | Charge | Charge |
| Molecular switch | Electron or <br> Atoms | Charge | Charge | Charge |
| NEMS | Atoms | Charge | Position | Charge |
| Atomic Switch | Atoms | Charge | Position | Electron |
| Memristor | Atoms | Charge | Charge, | Electron |
| Magnetic Cellular Automata | FM Domain | Magnetic dipole | Spin | FM Domain |
| Moving Domain Wall | FM Domain | Magnetic Dipole | Spin | FM Domain |
| Multi-Ferroic <br> Junction Tunnel | FM Domain | Spin | Charge | Electron |
| Optical or Plasmonics | Atoms <br> Electrons | Charge | Optical <br> Density | Photons |
| Thermal Transistor | Phonons | Thermal Energy | Temperature | Phonons |

Taxonomy for Candidate Information Processing Devices

## Breakthrough in Memristor

- non-volatile memories $\rightarrow$ low-power, high-density
- neuromorphic systems $\rightarrow$ Memristor mimics biological synapse
- As in a living creature the weight of a synapse is adapted by the ionic flow through it, so the conductance of a memristor is adjusted by the flux across or the charge through it depending on its controlling source.
- novel computer architectures $\rightarrow$ memory and process coexist
- Memristor will play a fundamental role in the realization of novel neuromorphic computing architectures merging memory and computation. This fundamental step will begin to bridge the main divide between biological computation and traditional computation, because memristor permits to bring data close to computation (the way biological systems do) and they use very little power to store that information.


## Fundamentals of memristor

- non-volatile memories
- neuromorphic systems
- computer architectures
$\rightarrow \quad$ low-power, high-density
$\rightarrow \quad$ memristors mimics synapses
$\rightarrow \quad$ memory and processing coexist

Important issues:

- full understanding of nonlinear dynamics
- modeling

Memristor modeling

## and

Nonlinear dynamics

## What is a MEMRISTOR?

## Memristor (L. O. Chua, 1971)

Memristor-The Missing Circuit Element

(a)
$f_{R}(v(t), i(t))=0$
$f_{C}(v(t), q(t))=0$
$f_{L}(\varphi(t), i(t))=0$
$f_{M}(\varphi(t), q(t))=0$

Charge-controlled "ideal" memristor
$\varphi(t)=f(q(t)) \Rightarrow v(t)=M(q(t)) i(t)$
(memristance) $M(q(t))=\frac{d f(q)}{d q}$

(b)

(c)


## General memristive one-port (Chua and Kang, 1976)

$$
v(t)=M(\mathbf{w}(t)) i(t), \quad \mathbf{w} \in R^{n}
$$

$$
\frac{d \mathbf{w}(t)}{d t}=h(\mathbf{w}(t), i(t), t)
$$

Main properties:

Recently, "Memristors" and "Memristive Devices" have been used interchangeably

■ passivity criterion $\Rightarrow M(\mathbf{w}(t)) \geq 0$
$\boxed{\square}$ non-volatile memory property $\Rightarrow h(\mathbf{w}(t), 0, t)=0, \quad \forall t$
■ v-i pinched hysteresis loop (Lissajous figure) for any periodic source. The pinched hysteresis loop shrinks continuously as the frequency increases


## General memristive one-port (Chua and Kang, 1976)

Identical zero-crossing
property


Fig. 5. Both $v(t)$ and $i(t)$ of a memristor with $0<R(\mathbf{x})<$ $\infty$ and $0<G(\mathbf{x})<\infty$ must have identical zero crossings.

Frequency-dependent pinched hysteresis loop property

Pinched hysteresis loop fingerprint


Example of a memristor pinched hysteresis loop.





Memory effects in complex materials and nanoscale systems

## HP Memristor (S. Williams et al, 2008)



## HP Memristor <br> (S. Williams et al, 2008)



Electrons: $\quad \nabla \cdot\left(-e n(x) \mu_{\mathrm{n}} \nabla \varphi_{\mathrm{n}}(x)\right)=0$

$$
\begin{array}{rc}
\text { Holes: } & \nabla \cdot\left(e p(x) \mu_{\mathrm{p}} \nabla \varphi_{\mathrm{p}}(x)\right)=0 \\
\text { Ions: } & -\nabla \cdot\left(-e D_{\mathrm{i}} \nabla N_{\mathrm{D}}(x)-e N_{\mathrm{D}}(x) \mu_{\mathrm{i}} \mathrm{E}_{0} \sinh \left[\nabla \varphi(x) / \mathrm{E}_{0}\right]\right)=e \partial N_{\mathrm{D}}(x) / \partial t \\
\text { Poisson: } & -\varepsilon \varepsilon_{0} \Delta \varphi(x)=e\left[p(x)-n(x)+f_{\mathrm{D}}(x) N_{\mathrm{D}}(x)-f_{\mathrm{A}}(x) N_{\mathrm{A}}\right]
\end{array}
$$

## Memristor mathematical model (HP model)



First memristor model

$$
\left\{\begin{array}{l}
\frac{d w(t)}{d t}=\mu \frac{R_{o n} \frac{w(t)}{D} i(t)}{w(t)} \\
v(t)=\left(R_{o n} \frac{w(t)}{D}+R_{o f f}\left(1-\frac{w(t)}{D}\right)\right) i(t)
\end{array}\right.
$$

Expression for $w(t)$ as function of $q(t)$ :

$$
w(t)=w\left(t_{0}\right)+\mu \frac{R_{o n}}{D}\left(q(t)-q\left(t_{0}\right)\right)
$$

For $R_{\text {on }} \ll R_{\text {off }}$ memristance expressed by

$$
M(q(t))=R_{\text {off }}\left(1-\frac{w\left(t_{0}\right)}{D}-\frac{\mu R_{o n}}{D^{2}}\left(q(t)-q\left(t_{0}\right)\right)\right)
$$

## Memristor mathematical model

$$
\begin{aligned}
i(t) & =W(x(t)) v(t) \\
W(x(t)) & =\frac{G_{o n} G_{o f f}}{G_{o n}-\Delta G x(t)} \quad \text { memductance }
\end{aligned}
$$

with $\Delta G=G_{o n}-G_{o f f}$, where $G_{o n}$ and $G_{o f f}$ indicate the device memductance respectively in the fully-conductive and fully-insulating state (i.e. as $x(t)$ sets to 1 and 0 respectively),

$$
\begin{array}{r}
\frac{d x(t)}{d t}=\frac{\eta}{i_{0}} W(x(t)) v(t) F(x(t), \eta v(t), p) \\
\text { window function }
\end{array}
$$

## Memristor mathematical model

- to account for nonlinear effects on the ionic transport Window funlction. $\begin{aligned} & \text { ( } p \in \mathbf{N}_{+} \text {modulates the degree of such nonlinearities); }\end{aligned}$
$(a): F_{J}, p=\{1,3,10\} \quad(b): F_{B}, p=2, \eta=1 \quad(c): \tilde{F}, \eta=1$




$$
\begin{aligned}
& F_{J}(x, p)=1-(2 x-1)^{2 p} \quad \forall \eta v \\
& F_{B}(x, \eta v, p)= \begin{cases}1-x^{2 p} & \eta v>0 \\
1-(x-1)^{2 p} & \eta v \leq 0\end{cases} \\
& F(x, v)=\left\{\begin{array}{l}
1 \text { if } C_{1} \text { holds }, \\
0 \text { if } C_{2} \text { or } C_{3} \text { holds. }
\end{array}\right. \\
& C_{1}=\left\{x \in(0,1) \text { or }\left(x=0 \text { and } v>v_{t h, 0}\right)\right. \\
& C_{2}=\left\{x=0 \text { and } v \leq v_{t h, 0}\right\}, \\
& C_{3}=\left\{x=1 \text { and } v \geq-v_{t h, 1}\right\},
\end{aligned}
$$

F. Corinto and A. Ascoli, 'A boundary condition-based approach to the modeling of memristor nano-structures', IEEE Trans. on Circ. and Syst.-I, 2012, DOI: 10.1109/ TCSI.2012.2190563

# Memristor mathematical model Boundary Condition-based Model (BCM) 

> [3] D. B. Strukov, G. S. Snider, D. R. Stewart, and R. S. Williams, "The missing memristor found," Nature, vol. 453, pp. 80-83, 2008.

> It may be theoretically demonstrated (see Proposition 3 in [1]) that boundary conditions may be tuned so as to obtain either singlevalued or multi-valued state-flux characteristics.


Figure 8. Memductance-flux characteristics for the memristor modeled by (1)-(2) $\left(x(0)=0.1, G_{o n}=10^{-2} S\right)$, using proposed window (39) and voltage source (6) with $\beta=0.01$. Blue curve: $v_{0}=1 V, G_{\text {on }} G_{\text {off }}^{-1}=60$, $v_{t h}=0 V$ (note the similarity to red curve in Fig. 4). Black curve: $v_{0}=2 V$, $G_{\text {on }} G_{o f f}^{-1}=25, v_{t h}=0 V$. Red curve: $v_{0}=1 V, G_{o n} G_{o f f}^{-1}=60$, $v_{t h}=0.5 v_{0} V$ (note the similarity to red curve in Fig. 2). For each case, points $P^{\prime}=\left(\varphi\left(t_{\alpha, 1}\right), G_{o n}\right)$ and $Q^{\prime}=\left(\varphi\left(t_{\beta}\right), G_{o n}\right)$ are highlighted to show whether or not one of conditions for single-valuedness, i.e. (45), is fulfilled. Here $P^{\prime} \equiv Q^{\prime}$ for the red curve only.

[^0]
# Memristor mathematical model Boundary Condition-based Model (BCM) 

[3] D. B. Strukov, G. S. Snider, D. R. Stewart, and R. S. Williams, "The missing memristor found," Nature, vol. 453, pp. 80-83, 2008.


Figure 9. Current-voltage characteristics corresponding to the memductanceflux relations of Fig. 8. Blue curve is similar to red curve in Fig. 5 and to Fig. 3(a) from [3]. Black curve is similar to Fig. 3(b) from [3]. Red curve is similar to red curve in Fig. 3 and to Fig. 3(c) from [3].

[^1]
# Memristor mathematical model Boundary Condition-based Model (BCM) 

## Table I

CAPABILITY OF EMULATION OF VARIOUS MEMRISTOR BEHAVIORS FROM [3] FOR VARIOUS MODELS PROPOSED IN LITERATURE.

| Memristor $i-v$ <br> response from [3] | Linear | Joglekar's | Biolek's | BCM |
| :--- | :---: | :---: | :---: | :---: |
| Fig. 2(b) in [3] | $\checkmark$ | $\checkmark$ | $\checkmark$ | Fig. 6 |
| Fig. 2(c) in [3] | $\checkmark$ | $\checkmark$ | $\checkmark$ | Fig. 7 |
| Fig. 3(a) in [3] | $X$ | $X$ | $\checkmark$ | Fig. 9 (blue) |
| Fig. 3(b) in [3] | $X$ | $X$ | $\checkmark$ | Fig. 9 (black) |
| Fig. 3(c) in [3] | $X$ | $\checkmark$ | $X$ | Fig. 9 (red) |

BCM model captures the $i-v$ response of other memristor nanostructures as well (see [1])
F. Corinto and A. Ascoli, 'A boundary condition-based approach to the modeling of memristor nano-structures', IEEE Trans. on Circ. and Syst.-I, 2012, DOI: 10.1109/TCSI.2012.2190563

## Memristor Model Comparison

$$
\begin{aligned}
\frac{d w}{d t}= & f_{\text {off }} \sinh \left(\frac{|i|}{i_{\text {off }}}\right) \\
& \exp \left(-\exp \left(\frac{w-a_{\text {off }}}{w_{c}}-\frac{|i|}{b}\right)-\frac{w}{w_{c}}\right)
\end{aligned}
$$

for $i>0$, while it is

$$
\begin{aligned}
& \frac{d w}{d t}=-f_{\text {on }} \sinh \left(\frac{|i|}{i_{\text {on }}}\right) \\
& \quad \exp \left(-\exp \left(\frac{a_{\text {on }}-w}{w_{c}}-\frac{|i|}{b}\right)-\frac{w}{w_{c}}\right)
\end{aligned}
$$

Alon Ascoli, Fernando Corinto, Vanessa singer, and Ronald Tetzlaff

| Table 4. <br> Comparison among the memristor models. For sake <br> of brevity we use the acronym BM to indicate Biolek's <br> Memristor. |
| :--- | BM $\quad$ BCM $\quad$ Team $\quad$| Test 1 | - | $\checkmark$ | - |
| :--- | :--- | :--- | :--- |
| Test 2-a | $\checkmark$ | $\checkmark$ | $\checkmark$ |
| Test 2-b | - | - | $\checkmark$ |
| Test 2-c | $\checkmark$ | $\checkmark$ | - |
| Test 3 | \# | \# | \# |

Figure: Test 1 aims to identify thich memristor model fits better the $i-v$ characteristic observed in the Pickett model under a particular triangular excitation. Test 2 memristor-based nonlinear circuit with periodic behavior (a) frequency of the limit cycle; b) frequency spectrum of the memristor voltage; c) transitory response due to an external pulse). Test 3 chaotic memristor-based non- linear circuit (appearance of the chaotic attractor).

## Memristor modeling

## Messagge to take home: Reliable mathematical and physicallybased circuit models are fundamental to develop neuromorphic hybrid systems!

Electrons:
Holes: Ions:
Poisson:

$$
\begin{aligned}
& \nabla \cdot\left(-e n(x) \mu_{\mathrm{n}} \nabla \varphi_{\mathrm{n}}(x)\right)=0 \\
& \nabla \cdot\left(e p(x) \mu_{\mathrm{D}} \nabla \varphi_{\mathrm{D}}(x)\right)=0 \\
& -\nabla \cdot\left(-e D_{\mathrm{i}} \nabla N_{\mathrm{D}}(x)-e N_{\mathrm{D}}(x) \mu_{\mathrm{i}} \mathrm{E}_{\mathrm{o}} \sinh \left[\nabla \varphi(x) / \mathrm{E}_{\mathrm{D}}\right]\right)=e \partial N_{\mathrm{D}}(x) / \partial t \\
& \quad-\varepsilon \varepsilon_{0} \Delta \varphi(x)=e\left[p(x)-n(x)+f_{\mathrm{D}}(x) N_{\mathrm{D}}(x)-f_{\mathrm{A}}(x) N_{\mathrm{A}}\right]
\end{aligned}
$$

$$
\begin{aligned}
v(t) & =M(x, i) i \\
\frac{d x(t)}{d t} & =f(x, i) \\
x(0) & =x_{0} \in R^{n}
\end{aligned}
$$

Nonlinear dynamics

## Memristor:

 The role of Initial ConditionsAnalysis of the memristive current-voltage behavior, i.e. make a rigorous classification of all possible current-voltage characteristics for a sine-wave voltage-driven memristive element on the basis of amplitude, angular frequency and time history of the oscillating voltage waveform across the device

## Memristor:



Resistance switching memories are memristors

Analysis of the memr ${ }^{\text {te }}$ behavior, i.e. make a of all possible currentfor a sine-wave volta element on the basis of amplitude, angular frequency and time history of the oscillating voltage waveform across the device

## Leon Chua

Abstract All 2-terminal non-volatile memory devices based on resistance switching are memristors, regardless of the device material and physical operating mechanisms. They all exhibit a distinctive "fingerprint" characterized by a pinched hysteresis loop confined to the first and the third quadrants of the $v-i$ plane whose contour shape in general changes with both the amplitude and frequency of any periodic "sine-wave-like" input voltage source, or current source. In par-

## Memristor: <br> The role of Initial Conditions

- voltage-controlled memristor



## Memristor: The role of Initial Conditions

| Flux initial value | Type of behavior |
| :---: | :--- |
| $\varphi(0) \geq \varphi_{c}$ | highly-resistive behavior |
| $\varphi_{c}-\frac{E}{\omega}<\varphi(0)<\varphi_{c}$ | atypical bow-tie with extended b-slope side <br> and cw rotation for $v>0$ |
| $\varphi(0)=\varphi_{c}-\frac{E}{\omega}$ | classical bow-tie with cw rotation for $v>0$ |
| $\varphi_{c}-\frac{2 E}{\omega}<\varphi(0)<\varphi_{c}-\frac{E}{\omega}$ | atypical bow-tie with extended a-slope side <br> and cw rotation for $v>0$ |
| $\varphi(0)=\varphi_{c}-\frac{2 E}{\omega}$ | limit-subcase 1, weakly-resistive behavior |
| $-\varphi_{c}<\varphi(0)<\varphi_{c}-\frac{2 E}{\omega}$ | weakly-resistive behavior |
| $\varphi(0)=-\varphi_{c}$ | limit-subcase 2, weakly-resistive behavior |
| $-\varphi_{c}-\frac{E}{\omega}<\varphi(0)<-\varphi_{c}$ | atypical bow-tie with extended a-slope side <br> and ccw rotation for $v>0$ |
| $\varphi(0)=-\varphi_{c}-\frac{E}{\omega}$ | classical bow-tie with ccw rotation for $v>0$ |
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## Memristor: The role of Initial Conditions



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| $\varphi(0)=-\varphi_{c}$ | limit-subcase 2, weakly-resistive behavior |
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## Memristor: The role of Initial Conditions



## Memristor: <br> The role of Initial Conditions

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| $\varphi(0)=\varphi_{c}-\frac{E}{\omega}$ | classical bow-tie with cw rotation for $v>0$ |
| $-\varphi_{c}<\varphi(0)<\varphi_{c}-\frac{E}{\omega}$ | atypical bow-tie with extended a-slope side and cw rotation for $v>0$ |
| $\varphi(0)=-\varphi_{c}$ | the only limit-case for $\frac{E}{\omega}=\varphi_{c}$, weaklyresistive behavior |
| $\varphi(0)=-\varphi_{c}$ | limit-subcase 1 for $\frac{E}{\omega} \neq \varphi_{c}$, atypical bow-tie with extended a-slope side and cw rotation for $v>0$ |
| $-\frac{E}{\omega}<\varphi(0)<-\varphi_{c}$ | cut atypical bow-tie with extended a-slope side and cw rotation for $v>0$ |
| $\varphi(0)=-\frac{E}{\omega}$ | limit-subcase 2 for $\frac{E}{\omega} \neq \varphi_{c}$, nonlinearlyresistive behavior |
| $\varphi_{c}-2 \frac{E}{\omega}<\varphi(0)<-\frac{E}{\omega}$ | cut atypical bow-tie with extended a-slope side and ccw rotation for $v>0$ |
| $\varphi(0)=\varphi_{c}-2 \frac{E}{\omega}$ | limit-subcase 3 for $\frac{E}{\omega} \neq \varphi_{c}$, atypical bow-tie with extended a-slope side and ccw rotation for $v>0$ |
| $-\varphi_{c}-\frac{E}{\omega}<\varphi(0)<\varphi_{c}-2 \frac{E}{\omega}$ | atypical bow-tie with extended a-slope side and ccw rotation for $v>0$ |
| $\varphi(0)=-\varphi_{c}-\frac{E}{\omega}$ | classical bow-tie with ccw rotation for $v>0$ |
| $-\varphi_{c}-\frac{2 E}{\omega}<\varphi(0)<-\varphi_{c}-\frac{E}{\omega}$ | atypical bow-tie with extended b-slope side and ccw rotation for $v>0$ |
| $\varphi(0) \leq-\varphi_{c}-\frac{2 E}{\omega}$ | highly-resistive behavior |


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| $\varphi(0)=-\varphi_{c}$ | limit-subcase 1 for $\frac{E}{\omega}=2 \varphi_{c}$, classical bow-tie with $c w$ rotation for ${ }^{\omega}>0$ |
| $-\frac{E}{\omega}=-2 \varphi_{c}<\varphi(0)<\varphi_{c}-\frac{E}{\omega}=-\varphi_{c}$ | cut atypical bow-tie with extended a-slope side and cw rotation for $v>0$ |
| $\varphi(0)=-\frac{E}{\omega}=-2 \varphi_{c}$ | limit-subcase 2 for $\frac{E}{\omega}=2 \varphi_{c}$, nonlinearlyresistive behavior |
| $-\varphi_{c}-\frac{E}{\omega}=-3 \varphi_{c}<\varphi(0)<-\frac{E}{\omega}=-2 \varphi_{c}$ | cut atypical bow-tie with extended a-slope side and ccw rotation for $v>0$ |
| $\varphi(0)=\varphi_{c}-\frac{2 E}{\omega}=-3 \varphi_{c}$ | limit-subcase 3 for $\frac{E}{\omega}=2 \varphi_{c}$, classical bow-tie with ccw rotation for $v>0$ |
| $\varphi(0)=-\varphi_{c}$ | limit-subcase 1 for $\frac{E}{\omega} \neq 2 \varphi_{c}$, atypical bow-tie with extended b -slope side and cw rotation for $v>0$ |
| $\varphi_{c}-\frac{E}{\omega}<\varphi(0)<-\varphi_{c}$ | cut atypical bow-tie with extended $b$-slope side and cw rotation for $v>0$ |
| $\varphi(0)=\varphi_{c}-\frac{E}{\omega}$ | cut classical bow-tie, cw rotation for $v>0$ |
| $-\frac{E}{\omega}<\varphi(0)<\varphi_{c}-\frac{E}{\omega}$ | cut atypical bow-tie with extended a-slope side and cw rotation for $v>0$ |
| $\varphi(0)=-\frac{E}{\omega}$ | limit-subcase 2 for $\frac{E}{\omega} \neq 2 \varphi_{c}$, nonlinearlyresistive behavior |
| $-\varphi_{c}-\frac{E}{\omega}<\varphi(0)<-\frac{E}{\omega}$ | cut atypical bow-tie with extended a-slope side and ccw rotation for $v>0$ |
| $\varphi(0)=-\varphi_{c}-\frac{E}{\omega}$ | cut classical bow-tie with ccw rotation for $v>$ 0 |
| $\varphi_{c}-\frac{2 E}{\omega}<\varphi(0)<-\varphi_{c}-\frac{E}{\omega}$ | cut atypical bow-tie with extended $b$-slope side, ccw rotation for $v>0$ |
| $\varphi(0)=\varphi_{c}-\frac{2 E}{\omega}$ | limit-subcase 3 for $\frac{E}{\omega} \neq 2 \varphi_{c}$, atypical bowtie with extended $b$-slope side, ccw rotation for $v>0$ |
| $-\varphi_{c}-\frac{2 E}{\omega}<\varphi(0)<\varphi_{c}-2 \frac{E}{\omega}$ | atypical bow-tie with extended b-slope side and ccw rotation for $v>0$ |
| $\varphi(0) \leq-\varphi_{c}-\frac{2 E}{\omega}$ | highly-resistive behavior |

F. Corinto, A. Ascoli, M. Gilli, "Analysis of Current-Voltage Characteristics for Memristive elements in Pattern Recognition Systems", Int. J. Circ. Th. and Appl. (2012)

## Memristor: The role of Initial Conditions



Type of behavior




$$
\begin{array}{|c|l|}
-\psi_{c}-\bar{\omega}<\psi(\cup)<-\psi_{c}-\bar{\omega} & \begin{array}{l}
\text { auypıai vuw-ue wiun exuenueu } \\
\text { and ccw rotation for } v>0
\end{array} \\
\hline \varphi(0) \leq-\varphi_{c}-\frac{2 E}{\omega} & \text { highly-resistive behavior siue } \\
\hline
\end{array}
$$


F. Corinto, A. Ascoli, M. Gilli, "Analysis of Current-Voltage Characteristics for Memristive elements in

Pattern Recognition Systems", Int. J. Circ. Th. and Appl. (2012)

## Memristor: <br> The role of Initial Conditions



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F. Corinto, A. Ascoli, M. Gilli, "Analysis of Current-Voltage Characteristics for Memristive elements in

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## Memristor: The role of Initial Conditions


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## Memristor: The role of Initial Conditions



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## Memristor:

## The role of nonlinear dynamics

Messagge to take home:
Be carefull in applications!
The $i$-v curve observed in memristor devices depends on the INPUT and the INITIAL CONDITION as well.

$$
\begin{aligned}
v(t) & =M(x, i) i \\
\frac{d x(t)}{d t} & =f(x, i) \\
x(0) & =x_{0} \in R^{n}
\end{aligned}
$$

## Memristor oscillators

Corinto, Ascoli and Gilli "Nonlinear dynamics of memristor oscillators", IEEE Trans. on Circuits and


$$
i_{m}=W\left(\varphi_{m}\right) v=\frac{d q_{m}\left(\varphi_{m}\right)}{d \varphi_{m}} \frac{d \varphi_{m}}{d t}
$$

## $W\left(\varphi_{m}\right)$ : memory conductance

## Memristor oscillator

With $R_{1}<0$ let

$$
\alpha=C_{1}^{-1}, \beta=-R_{1} L_{1}^{-1}, \xi=L_{1}^{-1}, \quad x_{1}=\hat{\varphi}_{1}, \quad x_{2}=\tilde{q}_{1}
$$

State equations are:

$$
\left\{\begin{array}{l}
\frac{d}{d t} x_{1}=\alpha x_{2}-\alpha \boldsymbol{q}_{m}\left(x_{1}\right) \\
\frac{d}{d t} x_{2}=-\xi x_{1}+\beta x_{2}
\end{array}\right.
$$

where

$$
q_{m}\left(x_{1}\right)=b x_{1}+(a-b) n\left(x_{1}\right), \quad n\left(x_{1}\right)=\frac{1}{2}\left(\left|x_{1}+1\right|-\left|x_{1}-1\right|\right)
$$

## Memristor oscillators

Corinto, Ascoli and Gilli "Nonlinear dynamics of memristor oscillators", IEEE Trans. on Circuits and
Systems-I, vol. 58, no. 6, pp. 1323-1336, 2011


Isoline $H b(a, b)=0$ of the surface for $\alpha=1.25$ and $\beta=\xi=1$ (blue curve)
In blue: isoline point $(a, b)=(0.5568,1.5)$

## Memristor chaotic circuits

> anti - parallel combination of memristive elements


$$
\frac{d x_{5}}{d \tilde{\tau}}=\frac{\eta_{2}}{\tilde{i}_{0}} W\left(x_{5}\right) x_{1} F\left(x_{5}, \eta_{2} W\left(x_{5}\right) x_{1}, p\right)
$$

$(a): \tilde{F}, v_{t h}=0$

$(b): F_{B}, p=10$

Setting circuit element values to $G=3.3 \mathrm{mS}, G_{N 1}=$ $-0.4 m S, G_{N 2}=-1.2 m S, C_{1}=50 n F, C_{2}=37 n F$ and $L=100 \mathrm{mH}$, system parameters are numerically given by $\tilde{\alpha}=0.74, \tilde{\beta}=0.0333$ and $\gamma=0.12$. Further $G_{o f f}=0.06 \mathrm{mS}$ and $G_{o n}=1.9 \mathrm{mS}$.

## Memristor-based Hodgkin-Huxley circuit



Intracellular medium
Table 3. Symbol and relevant equations defining the Potassium Ion-Channel Memristor.


Table 4. Symbol and relevant equations defining the Sodium Ion-Channel Memristor.


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DOI: 10.1142/S021812741230011X
HODGKIN-HUXLEY AXON IS MADE OF MEMRISTORS

LEON CHUA
VALERY SBITNEV
HYONGSUK KIM

## Memristor-based Hodgkin-Huxley circuit

## Memristor-based neural circuits

Fernando Corinto
Sung-Mo, "Steve" Kang
Extracellular medium


Fig. 2. Novel implementation of a memristor based on a special class of (passive) nonlinear resistive two-port connected to a nonlinear dynamic one-port.


Fig. 3. Nonlinear resistive two-port connected so as to satisfy equations (5)-(7). The current in each bipole has the direction specified by the arrow and voltage is defined by the associated reference direction.

# Novel circuit implementation 

## Memristive diode bridge with LCR filter

F. Corinto and A. Ascoli

## MEMORABLE EXHIBITION

PAGE 824 Researchers in Italy have shown that a purely passive circuit, employing already-existing components, can exhibit memristive dynamics. The circuit is composed of an elementary diode bridge and an RLC series circuit, introducing nonlinearity and dynamical behaviour into


Simple electronic systems can exhibit memristive behaviour

## ELECTRONICS LETTERS 5th July 2012 Vol. 48 No. 14

## Novel circuit implementation



$$
i_{g}=\left(i_{L}+2 I_{S}\right) \tanh \left(\frac{v_{g}}{2 n V_{T}}\right)
$$

$x_{1}=v\left(V_{T}\right)^{-1}$ and $x_{2}=i_{L}\left(I_{S}\right)^{-1}$



$$
\left[\begin{array}{c}
\beta\left(x_{2}-\alpha x_{1}\right) \\
\gamma\left(u-x_{1}-2 \ln \left(\frac{x_{2}+2}{2 \exp \left(-\frac{u}{2 n}\right) \cosh \left(\frac{u}{2 n}\right)}\right)\right)
\end{array}\right]
$$

Fig. 3 Current-voltage characteristics observed in numerical simulations of the mathematical model of the proposed circuit for a sine-wave input with $f$ set to 10 (plot (a)), 100 (plot (b)) and 1000 Hz (plot (c)).
$D 1 N 4148$, i.e. $I_{S}=2.682 n A$ and $n=1.836$, while $V_{T}=25 \mathrm{mV}$ $R=1.5 K \Omega \quad C=4 \mu F$ and $L=2.5 \mu H$


## Potassium-Ion Channel memristor emulator



Figure: First-order voltage-controlled memristor emulator.


Figure: $i(t)-v(t)$ curve of an individual potassium channel with sinusoidal input. The red curve is obtained by integrating the HH equations, while the blue curve is the output from the PSpice implementation of the memristor circuit.

## Perspectives

Memristor circuit modeling and nonlinear dynamics in
hybrid neuromorphic circuit with memristor-based computing nanoscale device.

## Unconventional Computing Systems

## Conclusion

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\author{

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}

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